

2F1120 Spektrala transformer för Media Solutions to Steiglitz Chapter 7

Preface

This document contains solutions to selected problems from Ken Steiglitz's book: "A Digital Signal Processing Primer" published by Addison-Wesley. Refer to the book for the problem text.

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7.1

From the text book:

 $\langle \mathbf{v}, \mathbf{w} \rangle = v_x w_x + v_y w_y + v_z w_z$

Where v_x is the component of the vector **v** along the *x*-axis, and so on for the other terms. If we consider only two dimensions for simplicity, and we define θ_v as the angle between the vector **v** and the *x*-axis:

$$v_x = |\mathbf{v}| \cos \theta_v$$
$$v_y = |\mathbf{v}| \sin \theta_v$$

We have then:

$$\begin{aligned} \langle \mathbf{v}, \mathbf{w} \rangle &= v_x w_x + v_y w_y \\ &= |\mathbf{v}| \cos \theta_v |\mathbf{w}| \cos \theta_w + |\mathbf{v}| \sin \theta_v |\mathbf{w}| \sin \theta_w \\ &= |\mathbf{v}| |\mathbf{w}| (\cos \theta_v \cos \theta_w + \sin \theta_v \sin \theta_w) \\ &= |\mathbf{v}| |\mathbf{w}| \cos (\theta_v - \theta_w) = |\mathbf{v}| |\mathbf{w}| \cos \theta_{vw} \end{aligned}$$

As we wanted to prove.

7.2

From geometry we know that three points in a three dimensional space always lie on a plane. Since \mathbf{v} and \mathbf{w} start from the same point (the origin), we can always consider them as lying on a plane and derive the expression for $\langle \mathbf{v}, \mathbf{w} \rangle$ the same way as in the previous example.



Figure 1.

7.3

From the text book (Equation 3.1):

$$C_{n} = \frac{1}{T} \left(\int_{0}^{T/2} e^{-jn\omega_{0}t} dt - \int_{T/2}^{T} e^{-jn\omega_{0}t} dt \right)$$

$$= \frac{1}{T} \left[\frac{1}{-jn\omega_{0}} e^{-jn\omega_{0}t} \right]_{0}^{T/2} - \frac{1}{T} \left[\frac{1}{-jn\omega_{0}} e^{-jn\omega_{0}t} \right]_{T/2}^{T}$$

$$= \frac{1}{-jn\omega_{0}T} \left(e^{-jn\omega_{0}T/2} - 1 \right) - \frac{1}{-jn\omega_{0}T} \left(e^{-jn\omega_{0}T} - e^{-jn\omega_{0}T/2} \right)$$

And, imposing $\omega_0 = 2\pi/T$:

$$= \frac{1}{j2n\pi} \left(1 - e^{-jn\pi}\right) + \frac{1}{j2n\pi} \left(e^{-j2n\pi} - e^{-jn\pi}\right)$$
$$= \frac{1}{j2n\pi} \left(1 - 2e^{-jn\pi} + (e^{-jn\pi})^2\right)$$
$$= \frac{1}{j2n\pi} \left(1 - e^{-jn\pi}\right)^2$$

If n is even $e^{-j2k\pi} = 1$, and $C_n = 0$, $\forall n = 2k$. If n is odd, $e^{-j2k\pi} = -1$, and $C_n = \frac{4}{j2n\pi} = -\frac{2j}{n\pi}$, $\forall n = 2k - 1$.

7.4

The Fourier coefficients are obtained multiplying the signal y(t) with a sinusoidal function s(t) at a certain frequency and then integrating over time. If s(t) is a sinus (an odd function), and the function y(t) is even, then their product g(t) = y(t)s(t) will be odd, that is g(t) = -g(-t). Since the integral extends over time, to each contributions for t < 0 corresponds its opposite for t > 0. This means that the integral will be zero, and so the coefficient in the Fourier series.

The same reasoning applies in the case y(t) is odd, and s(t) is a cosine.

The function is depicted in Figure 1. Following the same procedure as in Exercise 7.3 we obtain:

$$C_n = \frac{1}{T} \int_0^\alpha e^{-jn\omega_0 t} dt$$
$$= \frac{1}{T} \left[\frac{1}{-jn\omega_0} e^{-jn\omega_0 t} \right]_0^\alpha$$
$$= \frac{1}{-jn\omega_0 T} \left(e^{-jn\omega_0 \alpha} - 1 \right)$$

And with $\omega_0 = 2\pi/T$:

$$= \frac{j}{2n\pi} \left(e^{-j2n\pi\alpha/T} - 1 \right)$$

Note that in this case the result is dependent on the ratio between α and T. If $\alpha = T/2$ we obtain the same result as in 7.4.

7.5